

**Example 7.38** (Simple RC network). Consider the resistor-capacitor (RC) network shown in Figure 7.24 with input  $v_1$  and output  $v_2$ . This system is LTI and can be characterized by a linear differential equation with constant coefficients. (a) Find the system function  $H$  of this system. (b) Determine whether the system is BIBO stable. (c) Determine the step response of the system.

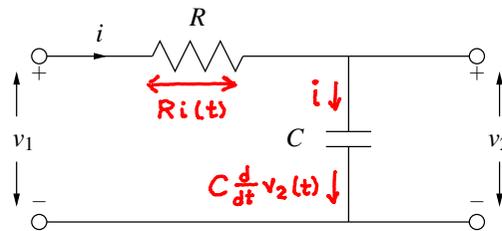


Figure 7.24: Simple RC network.

*Solution.* (a) From basic circuit analysis, we have

$$v_1(t) = Ri(t) + v_2(t) \quad \text{and} \quad (7.14a)$$

$$i(t) = C \frac{d}{dt} v_2(t). \quad (7.14b)$$

Taking the Laplace transform of (7.14) yields

$$V_1(s) = RI(s) + V_2(s) \quad \text{and} \quad (7.15a)$$

$$I(s) = CsV_2(s). \quad (7.15b)$$

Substituting (7.15b) into (7.15a) and rearranging, we obtain

$$V_1(s) = R[CsV_2(s)] + V_2(s)$$

$$\Rightarrow V_1(s) = RCsV_2(s) + V_2(s)$$

$$\Rightarrow V_1(s) = [1 + RCs]V_2(s)$$

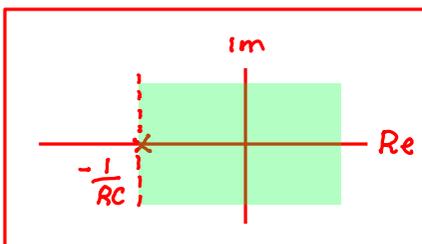
$$\Rightarrow \frac{V_2(s)}{V_1(s)} = \frac{1}{1 + RCs}$$

Thus, we have that the system function  $H$  is given by

$$H(s) = \frac{1}{1 + RCs}$$

$$= \frac{\frac{1}{RC}}{s + \frac{1}{RC}}$$

$$= \frac{\frac{1}{RC}}{s - (-\frac{1}{RC})}$$



Since the system can be physically realized, it must be causal. Therefore, the ROC of  $H$  must be a right-half plane. Thus, we may infer that the ROC of  $H$  is  $\text{Re}(s) > -\frac{1}{RC}$ . So, we have

$$H(s) = \frac{1}{1 + RCs} \quad \text{for } \text{Re}(s) > -\frac{1}{RC}.$$

(b) Since resistance and capacitance are (strictly) positive quantities,  $R > 0$  and  $C > 0$ . Thus,  $-\frac{1}{RC} < 0$ . Consequently, the ROC contains the imaginary axis and the system is stable.

(c) Now, let us calculate the step response of the system. We know that the system input-output behavior is characterized by the equation

$$\begin{aligned} V_2(s) &= H(s)V_1(s) \\ &= \left(\frac{1}{1+RCs}\right)V_1(s). \end{aligned}$$

← Since system is LTI  
← substitute for H

To compute the step response, we need to consider an input equal to the unit-step function. So,  $v_1 = u$ , implying that  $V_1(s) = \frac{1}{s}$ . Substituting this expression for  $V_1$  into the above expression for  $V_2$ , we have

$$\begin{aligned} V_2(s) &= \left(\frac{1}{1+RCs}\right)\left(\frac{1}{s}\right) \\ &= \frac{\frac{1}{RC}}{s\left(s+\frac{1}{RC}\right)}. \end{aligned}$$

$v_1(t) = u(t) \xleftrightarrow{\text{LT}} V_1(s) = \frac{1}{s}$   
← divide numerator and denominator by RC

Now, we need to compute the inverse Laplace transform of  $V_2$  in order to determine  $v_2$ . To simplify this task, we find the partial fraction expansion for  $V_2$ . We know that this expansion is of the form

$$V_2(s) = \frac{A_1}{s} + \frac{A_2}{s + \frac{1}{RC}}.$$

Solving for the coefficients of the expansion, we obtain

$$\begin{aligned} A_1 &= sV_2(s)|_{s=0} \\ &= 1 \quad \text{and} \\ A_2 &= \left(s + \frac{1}{RC}\right)V_2(s)|_{s=-\frac{1}{RC}} \\ &= \frac{\frac{1}{RC}}{-\frac{1}{RC}} \\ &= -1. \end{aligned}$$

Thus, we have that  $V_2$  has the partial fraction expansion given by

$$V_2(s) = \frac{1}{s} - \frac{1}{s + \frac{1}{RC}}.$$

Taking the inverse Laplace transform of both sides of the equation, we obtain

$$v_2(t) = \mathcal{L}^{-1}\left\{\frac{1}{s}\right\}(t) - \mathcal{L}^{-1}\left\{\frac{1}{s + \frac{1}{RC}}\right\}(t).$$

Using Table 7.2 and the fact that the system is causal (which implies the necessary ROC), we obtain

$$\begin{aligned} v_2(t) &= u(t) - e^{-t/(RC)}u(t) \\ &= \left(1 - e^{-t/(RC)}\right)u(t). \end{aligned}$$

inverse LT

$u(t) \xleftrightarrow{\text{LT}} \frac{1}{s} \text{ for } \text{Re}(s) > 0$  ■  
 $e^{-at}u(t) \xleftrightarrow{\text{LT}} \frac{1}{s+a} \text{ for } \text{Re}(s) > -a$